

Recent developments in wakefield computation



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Contents

- Wakefield simulation in the time domain
- CSR wakefield computations with high order DG
- Calculation of wakefields in the frequency domain
- Summary

* S. Schmid, *REPTIL - A relativistic 3D space charge particle tracking code based on the fast multipole method* (Wed, 9:15)

Wakefield simulation in the time domain

- Wakefield codes based on the solution of Maxwell's equations in the time domain

$$\frac{\partial B}{\partial t} = -\nabla \times E,$$

$$\frac{\partial \varepsilon E}{\partial t} = \nabla \times \frac{1}{\mu} B - j, \quad \boxed{j = qc\rho(x, y, z - ct)} \quad \text{rigid, ultra-relativistic bunch}$$

- Wake potentials and coupling impedances by post processing simulated field data

$$W_z(r_\perp, s) = -\frac{1}{q} \int_{-\infty}^{\infty} dz E_z \left(r_\perp, z, t = \frac{z + s}{c} \right)$$

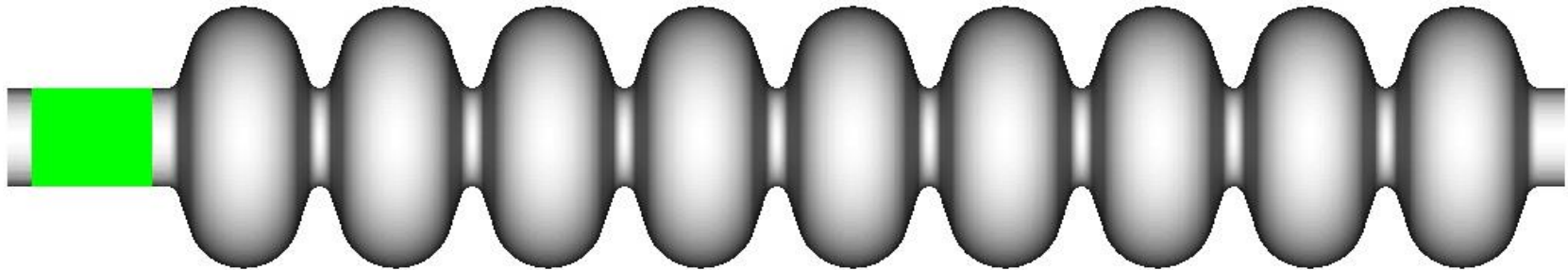
$$\frac{\partial}{\partial s} W_\perp(r_\perp, s) = -\nabla_\perp W_s(r_\perp, s)$$

Panofsky-Wenzel theorem

$$Z_z(r_\perp, \omega) = \frac{1}{c\rho(r_\perp, \omega)} \int_{-\infty}^{\infty} dz W_z(r_\perp, s) e^{-\frac{i\omega s}{c}}$$

The moving window

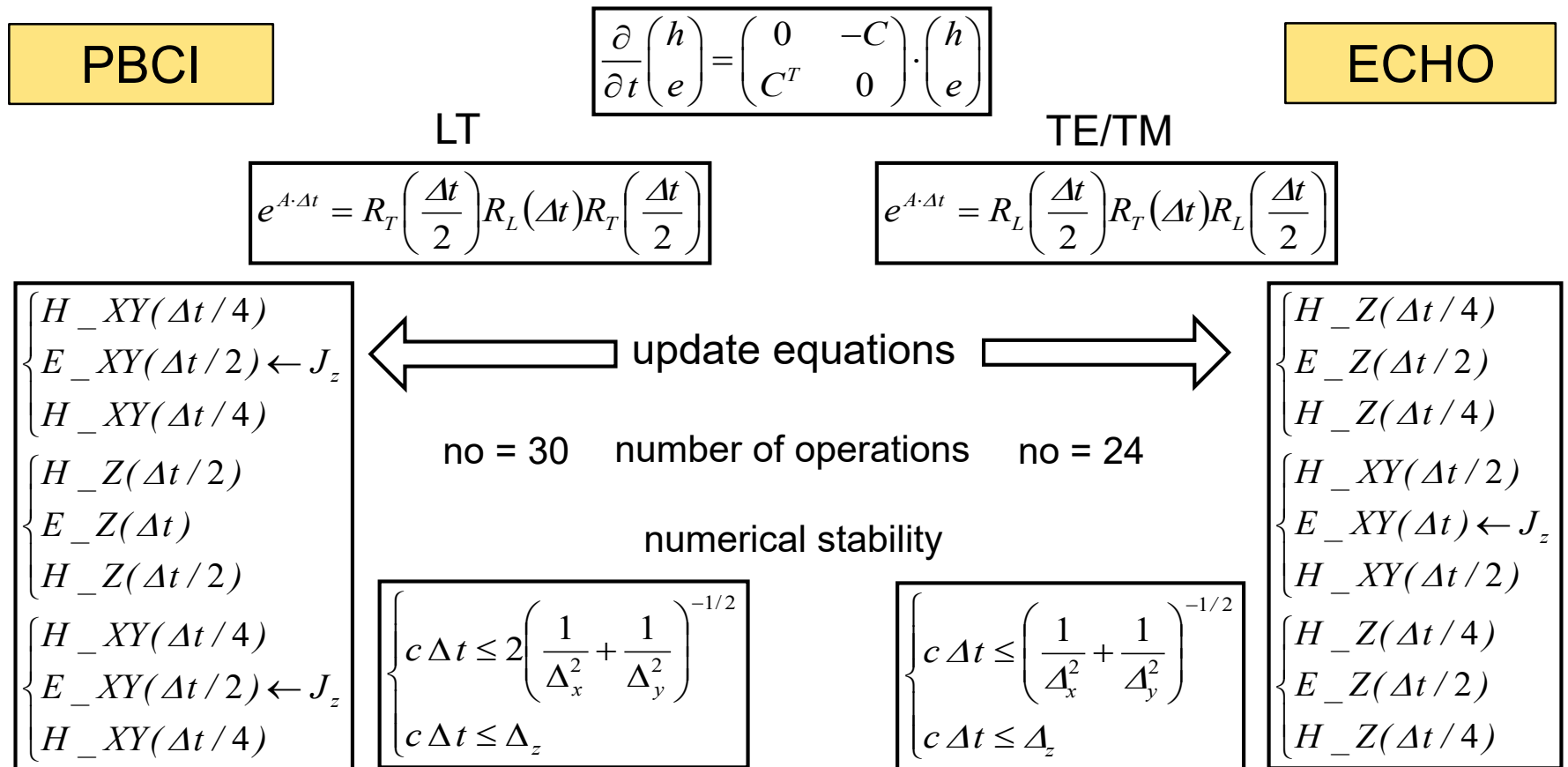
- Fields are calculated within a small computational frame which is co-moving with the bunch
 - Numerically efficient for large structures
 - Can handle complicated 3D-geometry by on-the-fly meshing
 - Can handle very short bunches



Wakefield (E_z) in a TESLA cavity computed with PBCI

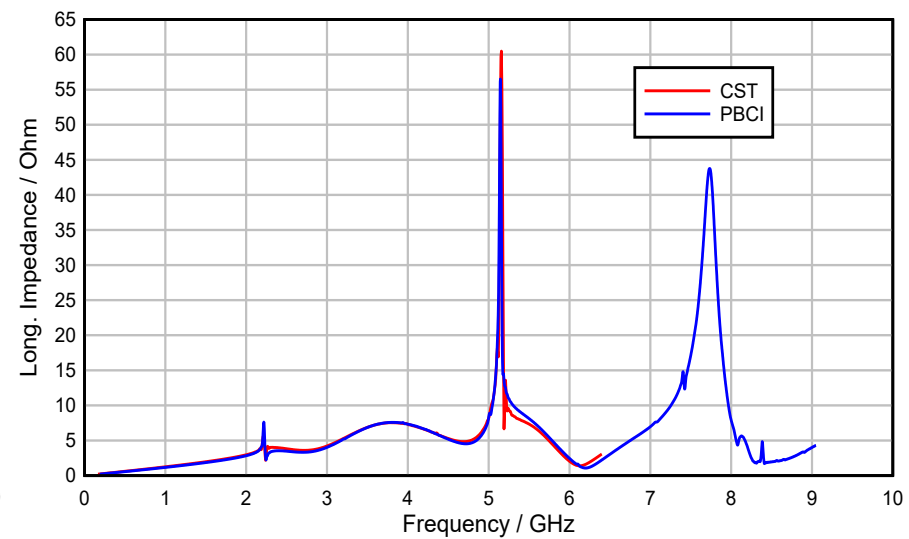
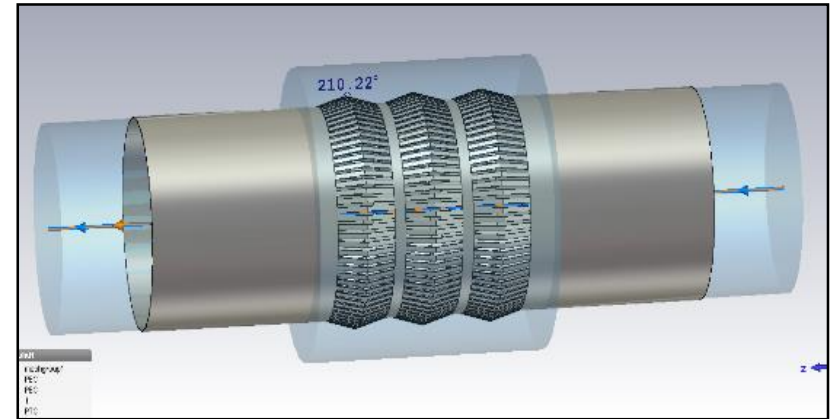
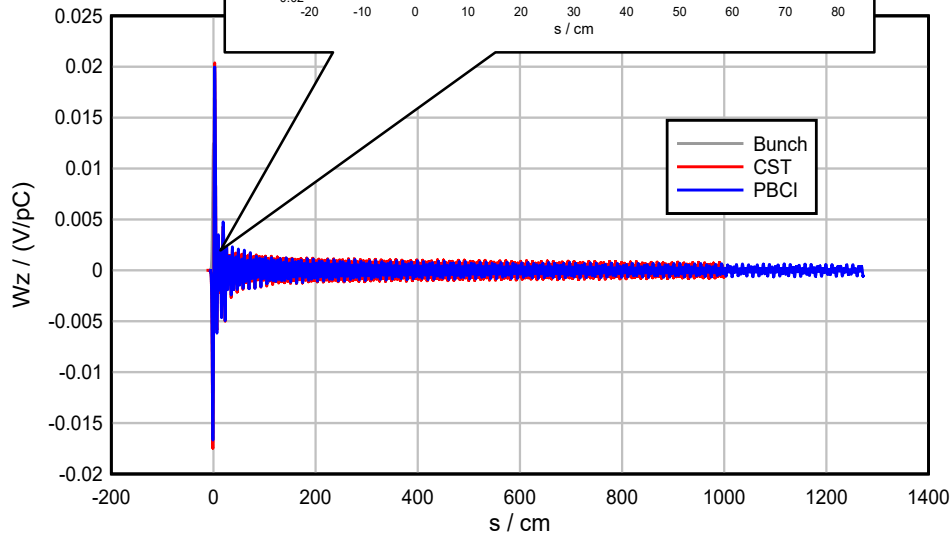
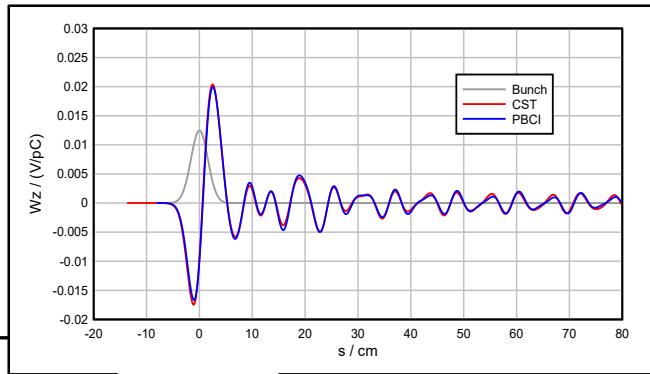
Dispersion-free methods

- Exact propagation in z-direction by splitting of the FDTD operators



Electrically large structures

- LHC RF-Fingers



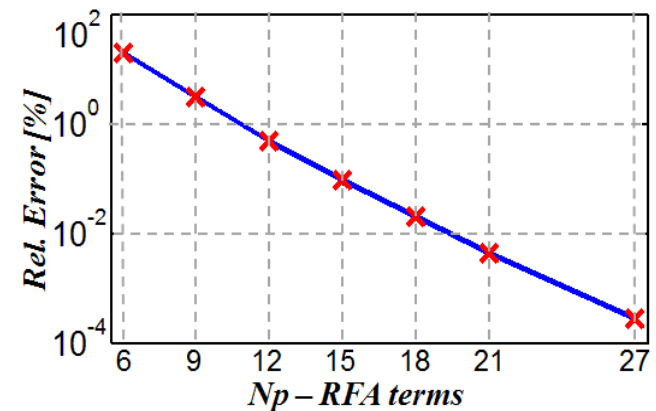
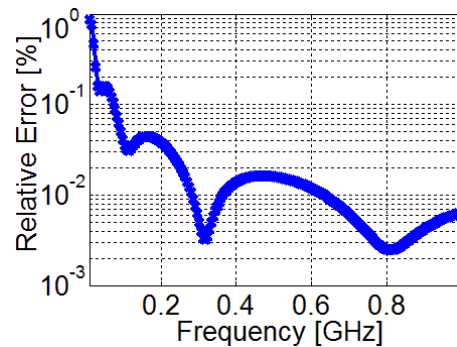
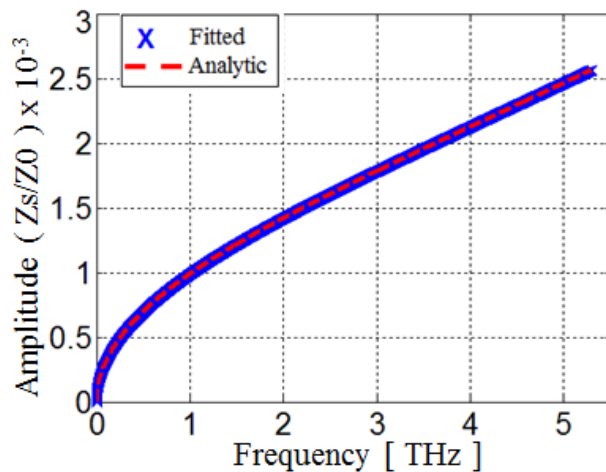
Resistive wall wakefields

- Generalized surface impedance functions

$$\vec{E}_\tau(\omega) = Z_s(\omega) [\vec{n} \times \vec{H}_\tau(\omega)] \quad \Rightarrow \quad Z_s(\omega) = j\omega L + \alpha_0 + \sum_{i=1}^{Np} \frac{\alpha_i}{j\omega + \beta_i} \quad \text{Pole-residue approximation}$$

- “Vector fitting” of resistive wall impedance: $Z_s(\omega) \cong \sqrt{\frac{j\omega\mu}{\sigma(\omega) + j\omega\varepsilon}}$

Example : Cu – N=21, ~ 10MHz-5THz, Δf~5MHz



- Auxiliary Differential Equation (ADE) formulation

$$\vec{n} \times \vec{E}(t) = L \cdot \frac{d}{dt} [\vec{n} \times \vec{n} \times \vec{H}(t)] + \sum_{i=0}^{Np} \vec{n} \times \vec{G}_i(t)$$

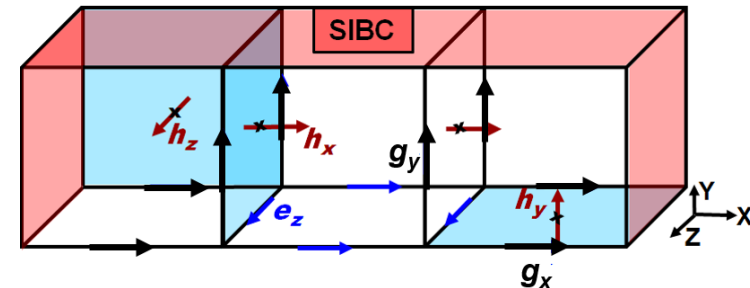
$$\vec{n} \times \vec{G}_0 = \alpha_0 [\vec{n} \times \vec{n} \times \vec{H}]$$

$$\frac{d}{dt} \vec{n} \times \vec{G}_i + \beta_i \vec{n} \times \vec{G}_i = \alpha_i [\vec{n} \times \vec{n} \times \vec{H}]$$

set of ADE for magnetic “surface currents”
(Woyna, Gjonaj, 2014)

- Modified discrete Maxwell’s equations:

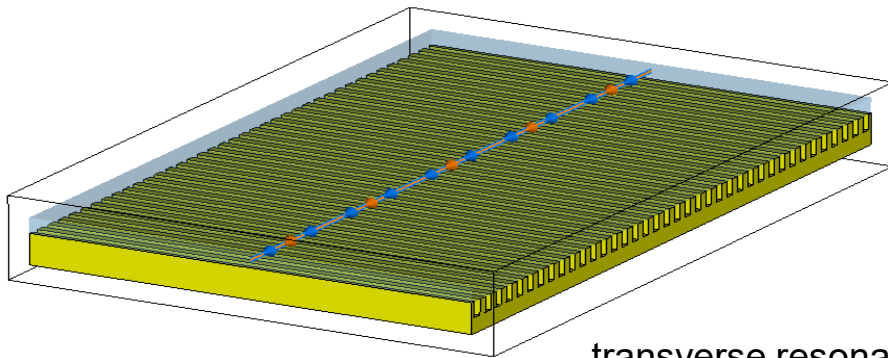
$$\frac{d}{dt} \begin{pmatrix} \hat{e} \\ \hat{h} \\ 0 \\ g_0 \\ \vdots \\ g_N \end{pmatrix} = \begin{pmatrix} 0 & M_\varepsilon^{-1} C^T & 0 & 0 & \dots & 0 \\ -M_\mu^{-1} C & 0 & C_B & C_B & \dots & C_B \\ 0 & \alpha_0 & 1 & 0 & \dots & 0 \\ 0 & -\alpha_1 & 0 & \beta_1 & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \ddots & \vdots \\ 0 & -\alpha_N & 0 & 0 & \dots & \beta_N \end{pmatrix} \begin{pmatrix} \hat{e} \\ \hat{h} \\ g_0 \\ g_1 \\ \vdots \\ g_N \end{pmatrix}$$



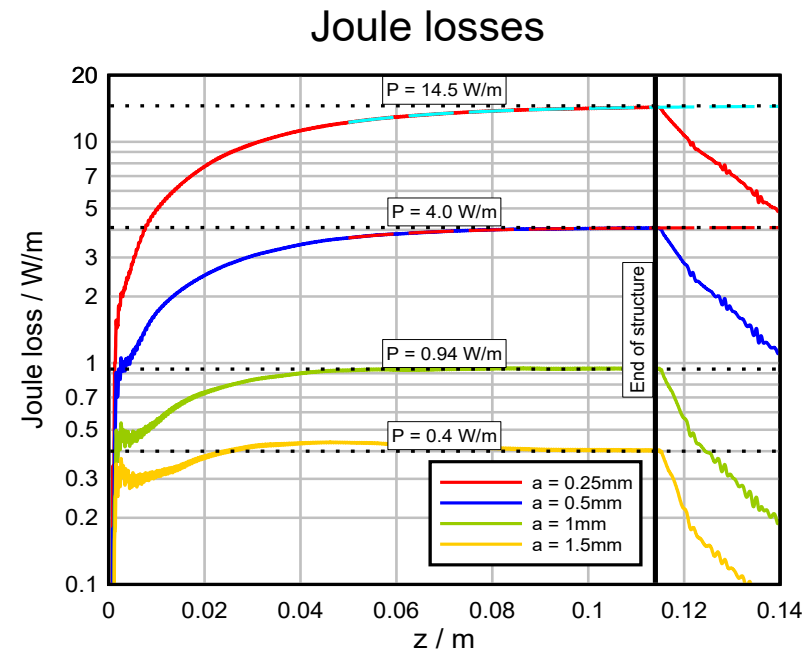
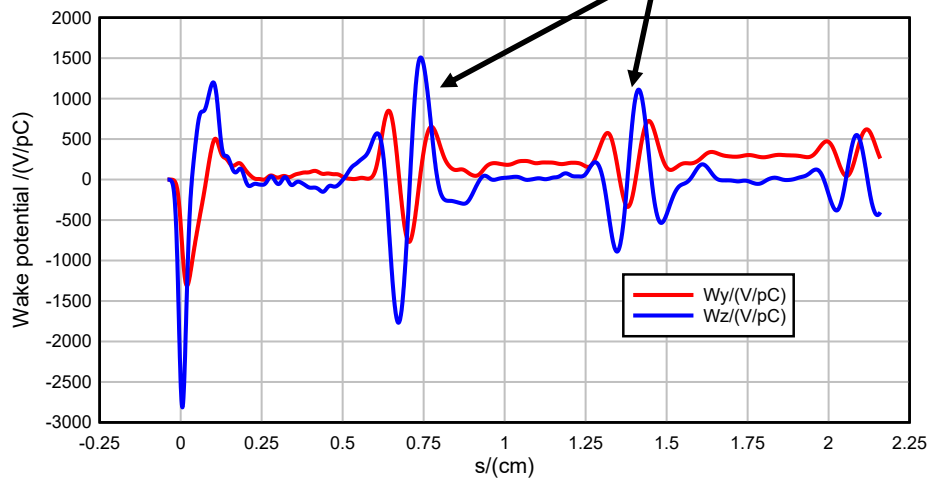
allocation of SIBC currents on grid

Open structures

- Single plate dechirper (with Bane, Stupakov)



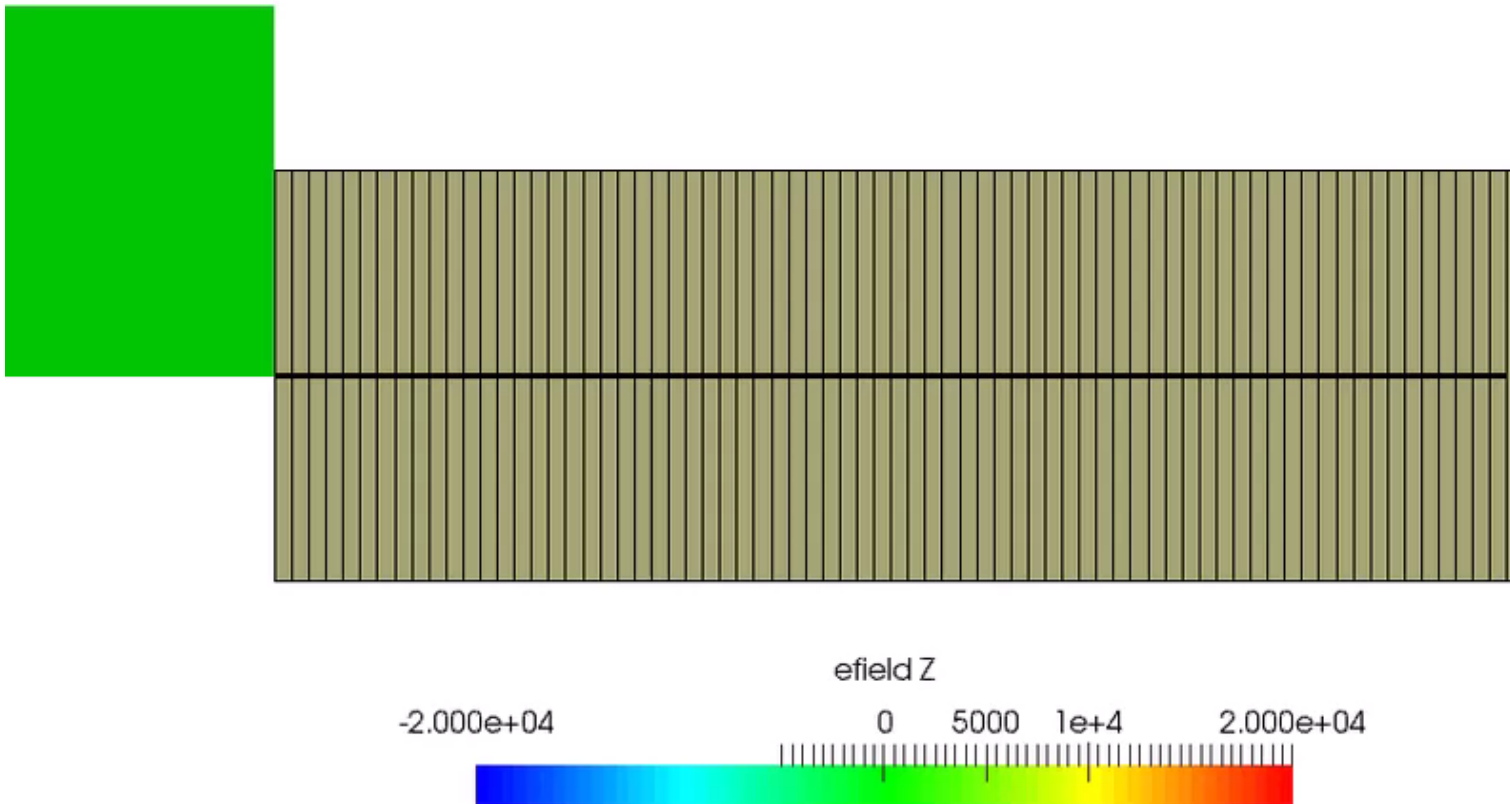
transverse resonances



Wake potentials for a 100um bunch

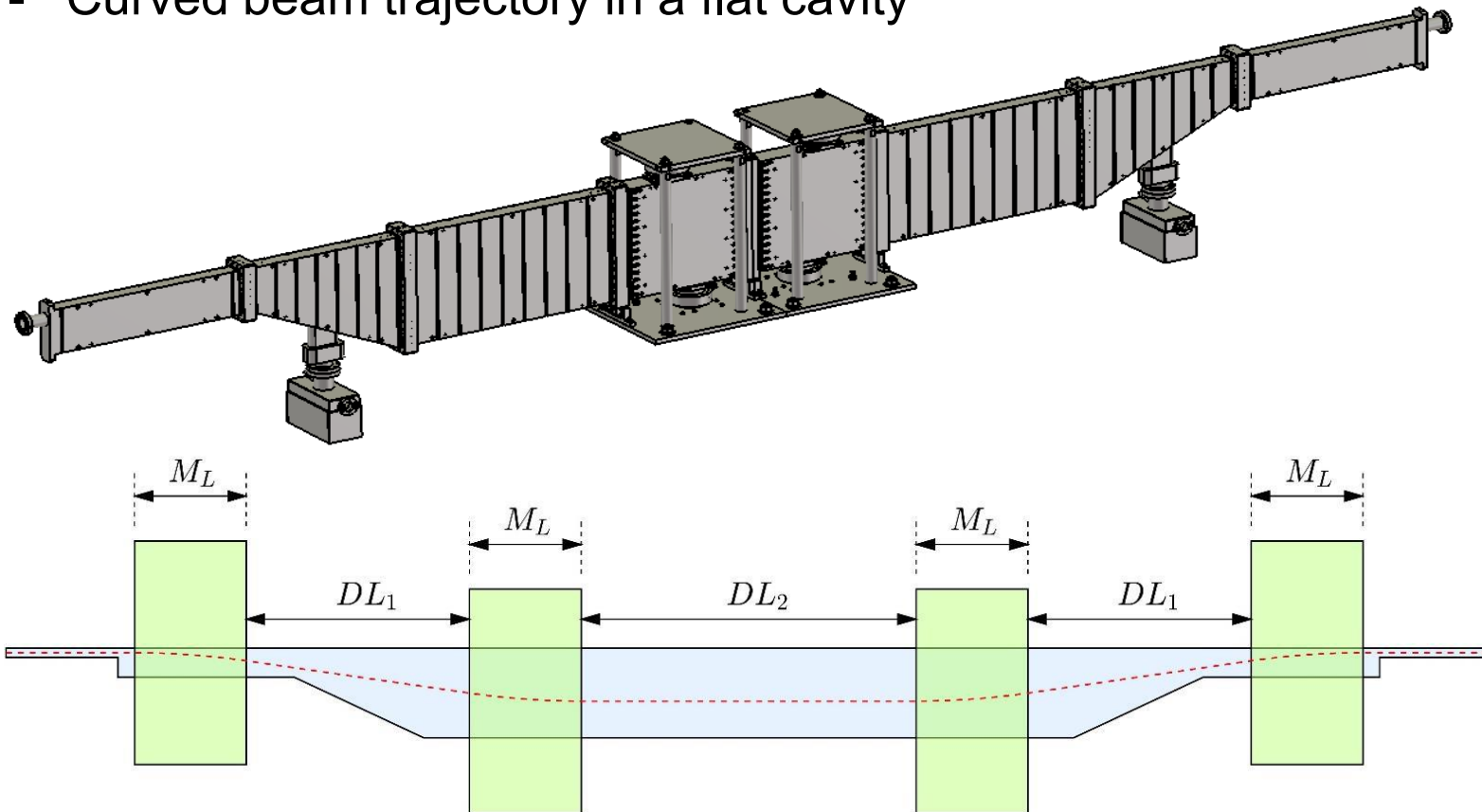
Open structures

- Single plate dechirper (with Bane, Stupakov)



CSR computations with high order DG

- DESY XFEL bunch compressor (BC0)
 - Curved beam trajectory in a flat cavity



CSR computations with high order DG

- CSRDG code (Bizzozero, Ellison, Warnock)
 1. Transform Maxwell's equations to Frenet-Serret coordinates
 2. Apply a Fourier mode decomposition in the y-direction

$$\frac{1}{Z_0} \frac{\partial E_{sp}}{\partial \tau} = \frac{\partial \tilde{H}_{yp}}{\partial x} + \alpha_p H_{xp}$$

$$\frac{1}{Z_0} \frac{\partial E_{xp}}{\partial \tau} = -\alpha_p H_{sp} - \frac{1}{\eta} \frac{\partial \tilde{H}_{yp}}{\partial s} - \frac{1}{\eta} q c G_p \lambda'(s - \tau) \Theta(x)$$

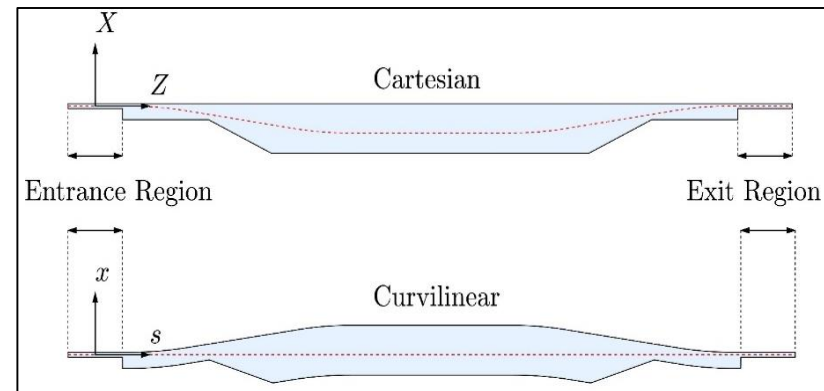
$$\frac{1}{Z_0} \frac{\partial E_{yp}}{\partial \tau} = \frac{1}{\eta} \frac{\partial H_{xp}}{\partial s} - \frac{\partial H_{sp}}{\partial x} - \frac{\kappa}{\eta} H_{sp}$$

$$Z_0 \frac{\partial H_{sp}}{\partial \tau} = \alpha_p E_{xp} - \frac{\partial E_{yp}}{\partial x}$$

$$Z_0 \frac{\partial H_{xp}}{\partial \tau} = \frac{1}{\eta} \frac{\partial E_{yp}}{\partial s} - \alpha_p E_{sp}$$

$$Z_0 \frac{\partial \tilde{H}_{yp}}{\partial \tau} = \frac{\partial E_{sp}}{\partial x} + \frac{\kappa}{\eta} E_{sp} - \frac{1}{\eta} \frac{\partial E_{xp}}{\partial s} + q Z_0 c G_p \lambda'(s - \tau) \Theta(x)$$

Transformation of geometry



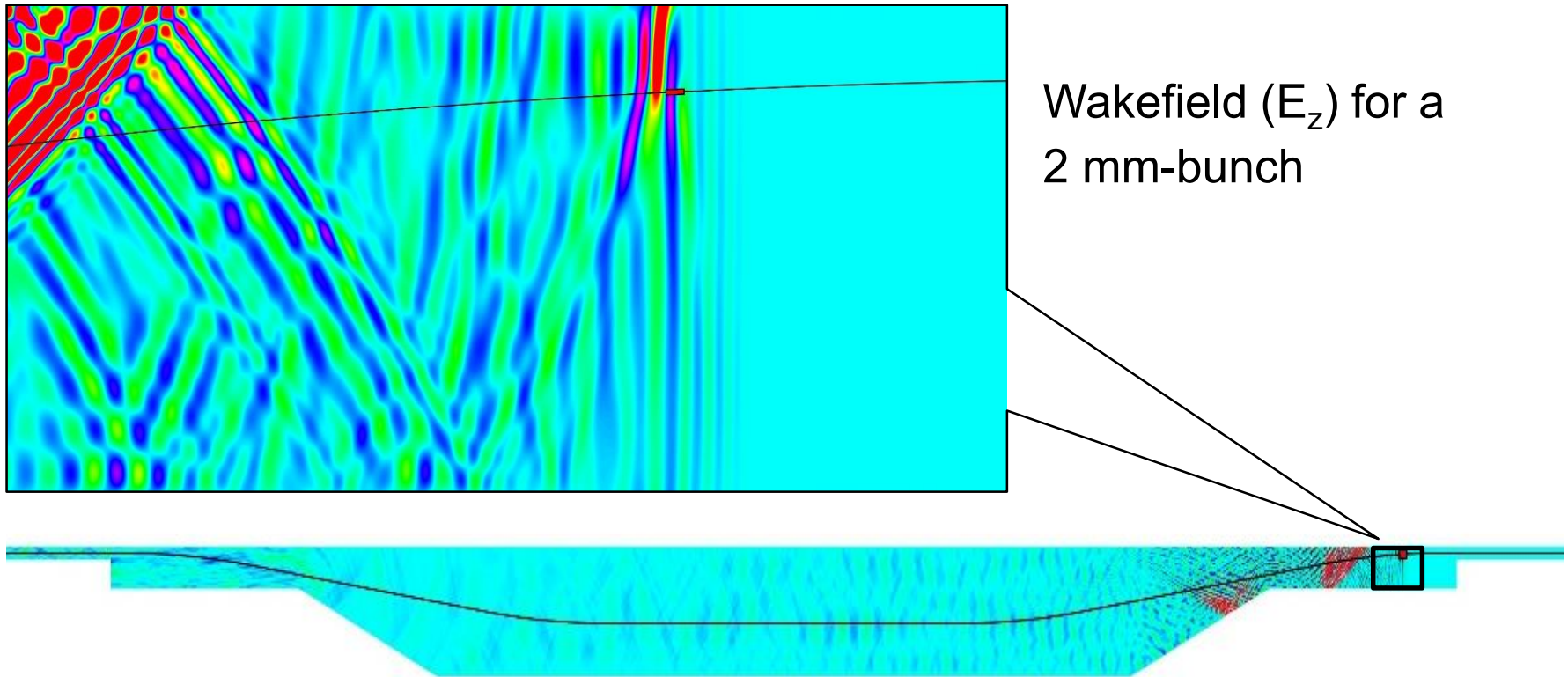
- CSRDG code (Bizzozero, Ellison, Warnock)
 3. Discretize the 2D-equations for each mode using the high order nodal discontinuous Galerkin method

$$\begin{aligned}
 \frac{dE_{sp}}{d\tau} &= Z_0 \mathcal{D}_x \tilde{H}_{yp} + Z_0 \alpha_p H_{xp} \\
 &\quad + \frac{1}{2} (J\mathcal{M})^{-1} \left(-Z_0 \mathbf{n}_x \llbracket \tilde{H}_{yp} \rrbracket - \llbracket E_{sp} \rrbracket + \mathbf{n}_s (\mathbf{n}_s \llbracket E_{sp} \rrbracket + \mathbf{n}_x \llbracket E_{xp} \rrbracket) \right) \\
 \frac{dE_{xp}}{d\tau} &= -Z_0 \alpha_p H_{sp} - \frac{Z_0}{1 + \kappa x} \mathcal{D}_s \tilde{H}_{yp} - \frac{Z_0}{1 + \kappa x} qcG_p \lambda'(s - \tau) \Theta(x) \\
 &\quad + \frac{1}{2} (J\mathcal{M})^{-1} \left(\frac{Z_0}{1 + \kappa x} \mathbf{n}_s \llbracket \tilde{H}_{yp} \rrbracket - \llbracket E_{xp} \rrbracket + \mathbf{n}_x (\mathbf{n}_s \llbracket E_{sp} \rrbracket + \mathbf{n}_x \llbracket E_{xp} \rrbracket) \right) \\
 \frac{dE_{yp}}{d\tau} &= \frac{Z_0}{1 + \kappa x} \mathcal{D}_s H_{xp} - Z_0 \mathcal{D}_x H_{sp} - \frac{Z_0 \kappa}{1 + \kappa x} H_{sp} \\
 &\quad + \frac{1}{2} (J\mathcal{M})^{-1} \left(-\frac{Z_0}{1 + \kappa x} \mathbf{n}_s \llbracket H_{xp} \rrbracket + Z_0 \mathbf{n}_x \llbracket H_{sp} \rrbracket - \llbracket E_{yp} \rrbracket \right)
 \end{aligned}$$

...and similarly for the magnetic components

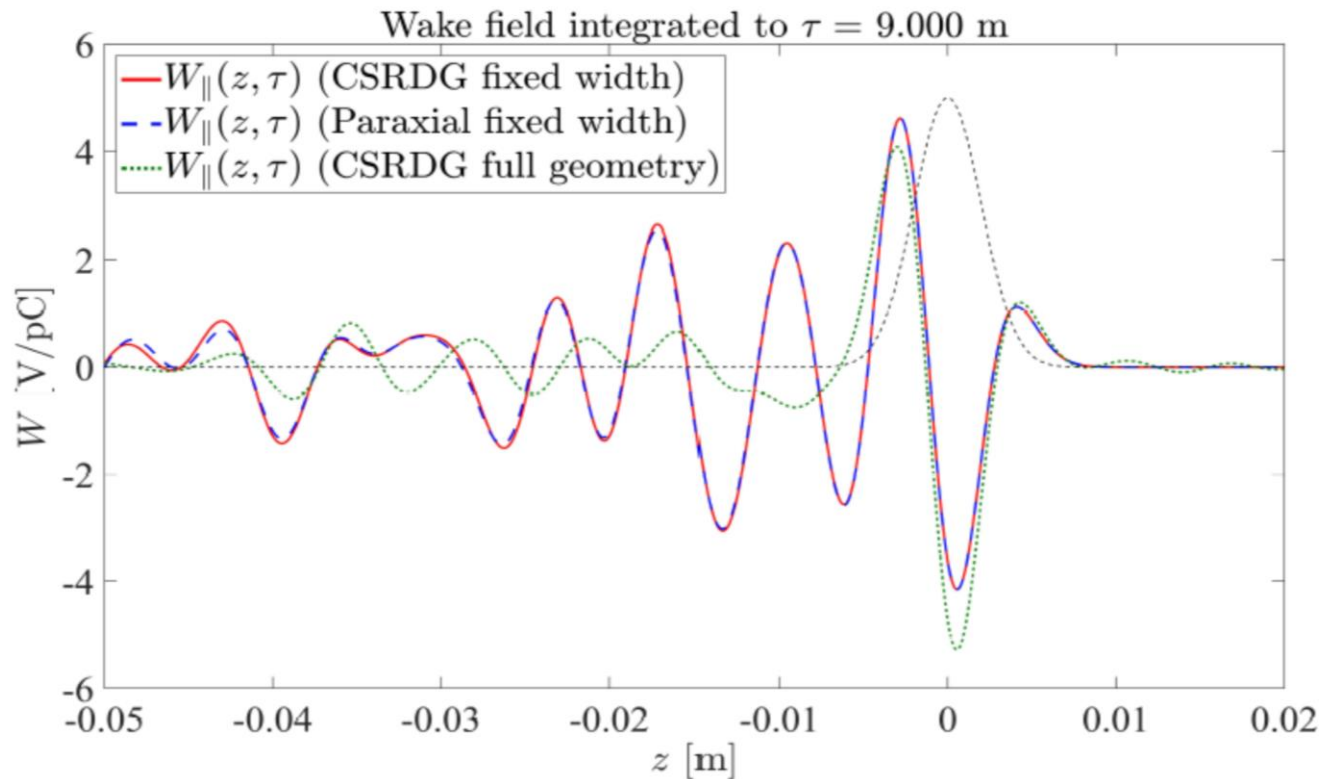
CSR computations with high order DG

- CSR DG code (Bizzozero, Ellison, Warnock)



CSR computations with high order DG

- CSRDG code (Bizzozero, Ellison, Warnock)



* D. Bizzozero, *Exploring the validity of the paraxial approximation for coherent synchrotron radiation wake fields* (Mo., 11:45)

Wakefields in the Frequency Domain

- Long range wakefields are needed
 - Low frequency, long bunches (ion, proton accelerators), bunch trains and/or high repetition rate
 - Wall heating (in resonant structures)
- Approximation of complicated geometry
 - Geometrical details smaller than bunch length
 - Smooth tapering etc. – Cartesian mesh approx. not sufficient
- Beams with $\beta < 1$
 - TD: Restriction in the choice of time step (M. Balk et al.)
 - TD: Expensive convolution for waveguide boundary conditions
- Curved beam trajectories
 - Moving window in the time domain not possible/complicated

Wakefields in the Frequency Domain

- The frequency domain problem

$$\nabla \times \mu^{-1} \nabla \times E - k_0^2 \varepsilon E = -jk_0 Z_0 J_s \quad J_s(x, y, z, \omega) = \rho(x, y) e^{-i \frac{\omega}{v} z}$$

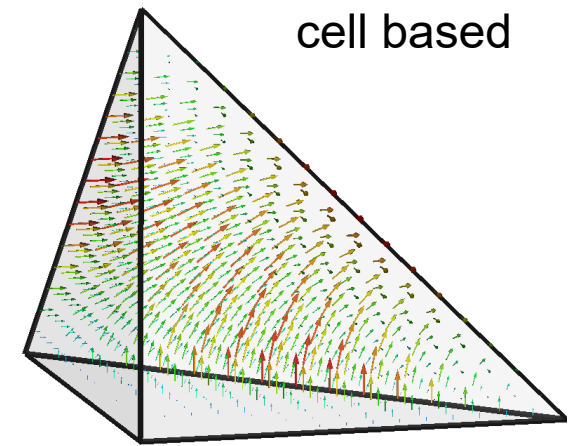
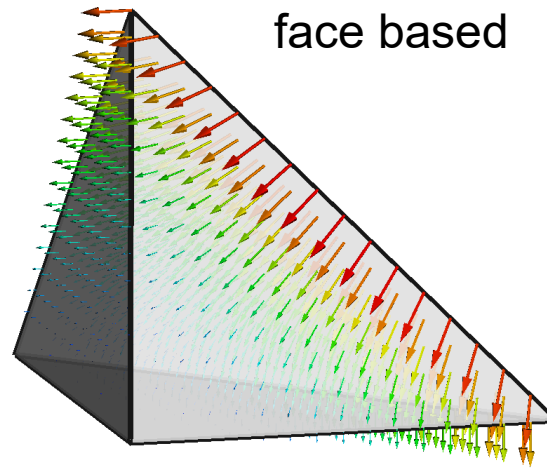
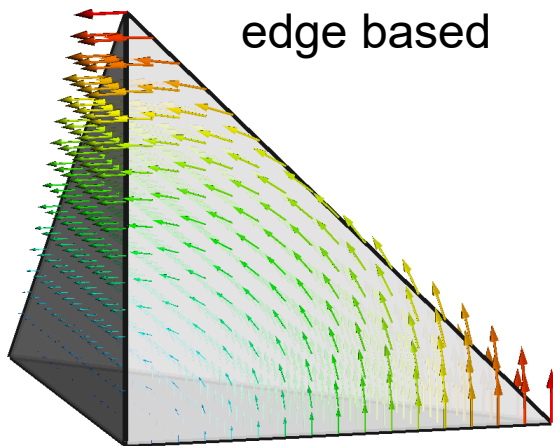
- Weak FE formulation: find $E \in H(\text{curl})$ such that:

$$\int dV \mu^{-1} \nabla \times E \cdot \nabla \times v_h - k_0^2 \int dV \varepsilon E \cdot v_h = -jk_0 Z_0 \int dV J_s \cdot v_h + \oint_S dS n \cdot [v_h \times \mu^{-1} \nabla \times E] \quad \forall v_h \in H(\text{curl})$$

$$\begin{array}{ccccccc} H_1 & \xrightarrow{\text{grad}} & H(\text{curl}) & \xrightarrow{\text{curl}} & H(\text{div}) & \xrightarrow{\text{div}} & L_2 \\ \cup & & \cup & & \cup & & \cup \\ W_h^{p+1} & \xrightarrow{\text{grad}} & v_h^p & \xrightarrow{\text{curl}} & q_h^{p-1} & \xrightarrow{\text{div}} & S_h^{p-2} \end{array}$$

Wakefields in the Frequency Domain

- High-order hierarchic basis functions*



- Allows for simple hp-adaption
- **Supports mesh elements of different type + hybrid meshes**

*M. Ainsworth, J. Coyle: *Int. J. of Numerical Methods in Eng.*, 2003.

*J. Schöberl, S. Zaglmayr: *Int. J. Comp. and Math. in Electrical and Electronic Eng.*, 2005.

Wakefields in the Frequency Domain

- Treatment of boundary surfaces

$$\int dV \mu^{-1} \nabla \times E \cdot \nabla \times v_h - k_0^2 \int dV \varepsilon E \cdot v_h =$$
$$-jk_0 Z_0 \int dV J_s \cdot v_h + \underbrace{\int_{S_{SIBC}} dS n \cdot [v_h \times \mu^{-1} \nabla \times E]}_{\text{resistive walls}} + \underbrace{\int_{S_{SWG}} dS n \cdot [v_h \times \mu^{-1} \nabla \times E]}_{\text{in \& outgoing pipes}}$$

- SIBC boundaries

$$\oint_{S_{SIBC}} dS n \cdot [v_h \times \mu^{-1} \nabla \times E] = \dots = j\omega \mathbf{Y}_S(\omega) \oint_{S_{SIBC}} dS v_h \cdot E$$

Simple modification of the system matrix on SIBC surfaces

No fitting of the surface impedance function or ADE/convolution is needed

Wakefields in the Frequency Domain

- Treatment of boundary surfaces

$$\int dV \mu^{-1} \nabla \times E \cdot \nabla \times v_h - k_0^2 \int dV \varepsilon E \cdot v_h =$$

$$-jk_0 Z_0 \int dV J_s \cdot v_h + \underbrace{\int_{S_{SIBC}} dS n \cdot [v_h \times \mu^{-1} \nabla \times E]}_{\text{resistive walls}} + \underbrace{\int_{S_{SWG}} dS n \cdot [v_h \times \mu^{-1} \nabla \times E]}_{\text{in \& outgoing pipes}}$$

- Beam pipe boundaries

$$n \times \nabla \times E = n \times \nabla \times E^{inc} + \sum_m a_m^{TE} \gamma_m^{TE} e_m^{TE} + \sum_m a_m^{TM} \gamma_m^{TM} e_m^{TM}$$

$$a_m^{TE} = \int_{S_{SWG}} dS e_m^{TE} \cdot [E - E^{inc}]$$

Reflection coefficients for each mode

$$a_m^{TM} = \int_{S_{SWG}} dS e_m^{TM} \cdot [E - E^{inc}]$$

Wakefields in the Frequency Domain

- Beam pipe boundary conditions

$$\int dV \mu^{-1} \nabla \times E \cdot \nabla \times v_h - k_0^2 \int dV \varepsilon E \cdot v_h + \sum_m P_m^{TE}(E) + \sum_m P_m^{TM}(E) =$$

$$-jk_0 Z_0 \int dV J_s \cdot v_h + \oint_{S_{WG}} dS \mathbf{n} \cdot [\mathbf{v}_h \times \mu^{-1} \nabla \times E^{inc}] + \sum_m U_m^{TE} + \sum_m U_m^{TM}$$

with $P_m^{TE}(E) = -\gamma_m^{TE} \left(\int_{S_{WG}} dS v_h \cdot e_m^{TE} \right) \left(\int_{S_{WG}} dS e_m^{TE} \cdot E \right)$, $P_m^{TM}(E) = \dots$

and matrix representation (TE):

$$P_m^{TE}(E) \rightarrow \mathbf{P}_m^{TE} \cdot \mathbf{e} = -\gamma_m^{TE} \mathbf{R}^T \cdot \mathbf{M}_m^{TE} \cdot \mathbf{R} \cdot \mathbf{e}$$

$$[\mathbf{R}]_{ij} = \int_{S_{WG}} dS \varphi_i^{2D} \cdot \varphi_j^{3D}$$

$$\mathbf{M}_m^{TE} = \mathbf{e}_m^{TE} \otimes \mathbf{e}_m^{TE} \quad \text{small but dense modal dyadic}$$

3D-to-2D projection matrix

Wakefields in the Frequency Domain

- Beam pipe boundary excitation

- For an ultra-relativistic bunch (same idea for $\beta < 1$):

$$\nabla_t \cdot E^{inc} = \frac{1}{\epsilon_0} \rho(x, y) e^{-ik_0 z_0}$$

$$\nabla \times E^{inc} = 0$$



2D-electrostatic problem at both ends of the pipe

- Modal contribution to the RHS

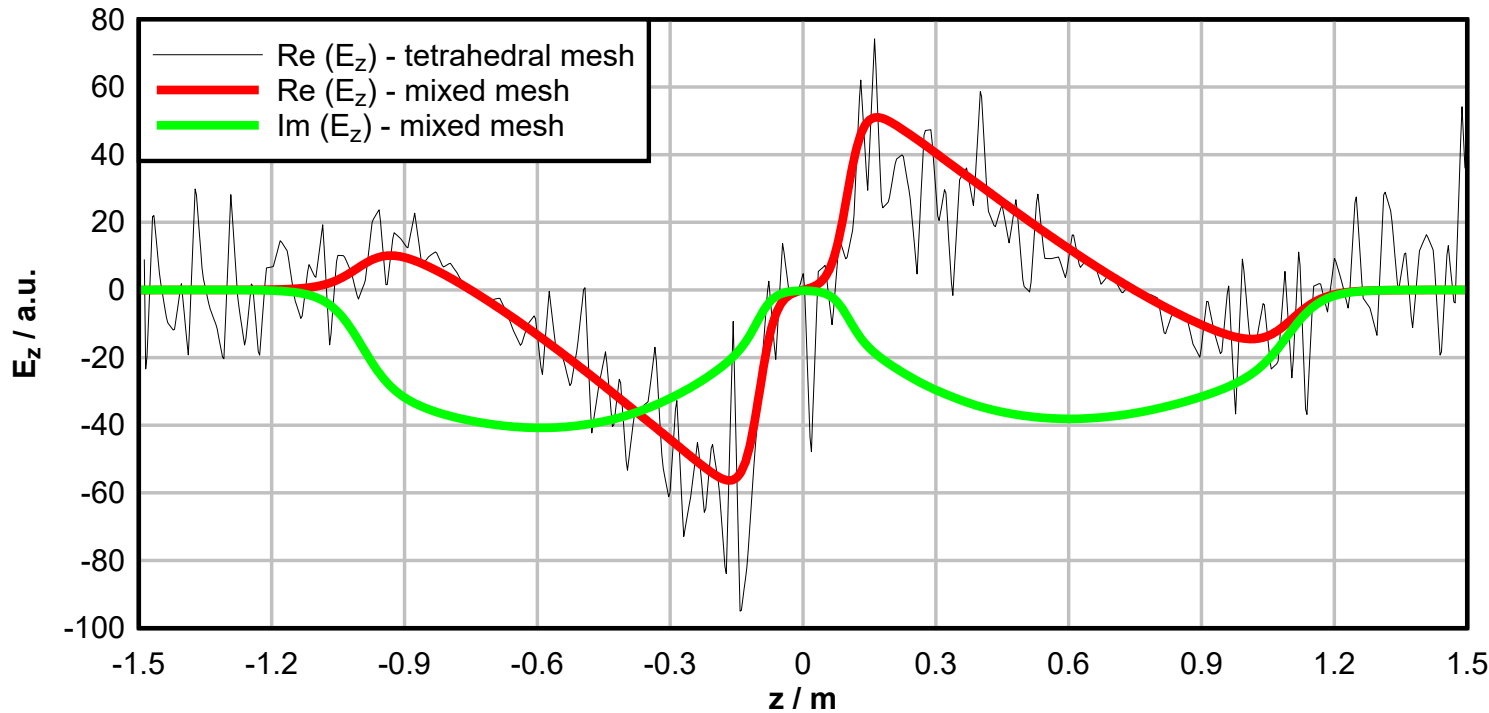
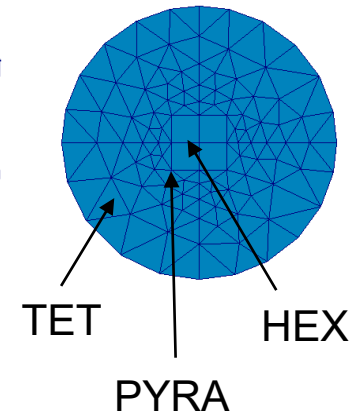
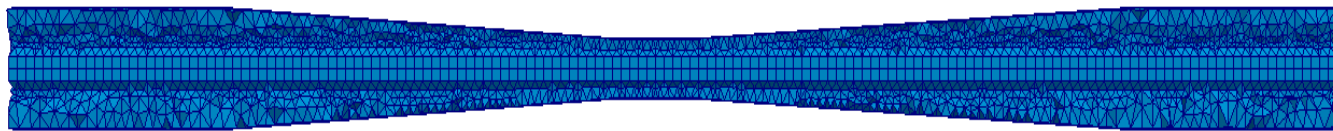
$$U_m^{TE}(E^{inc}) = -\gamma_m^{TE} \left(\int_{S_{WG}} dS v_h \cdot e_m^{TE} \right) \left(\int_{S_{WG}} dS e_m^{TE} \cdot E^{inc} \right)$$

$$U_m^{TE}(E^{inc}) \rightarrow \mathbf{U}_m^{TE} \cdot \mathbf{e}^{inc} = -\gamma_0^{TE} \mathbf{R}^T \cdot \mathbf{M}_m^{TE} \cdot \mathbf{R}^{2D} \cdot \mathbf{e}^{inc}$$

...for all waveguide modes supported in the pipe

Wakefields in the Frequency Domain

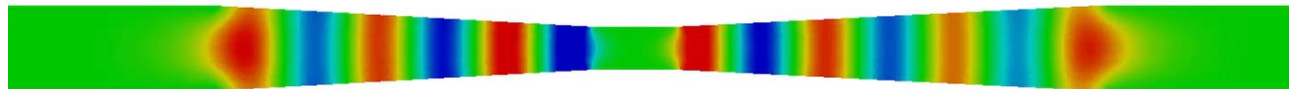
- Collimator example – use of hybrid meshes



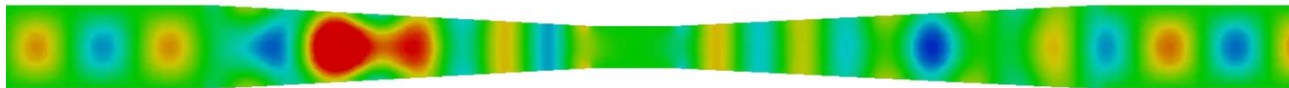
Longitudinal
wakefield on axis
at 100MHz

Wakefields in the Frequency Domain

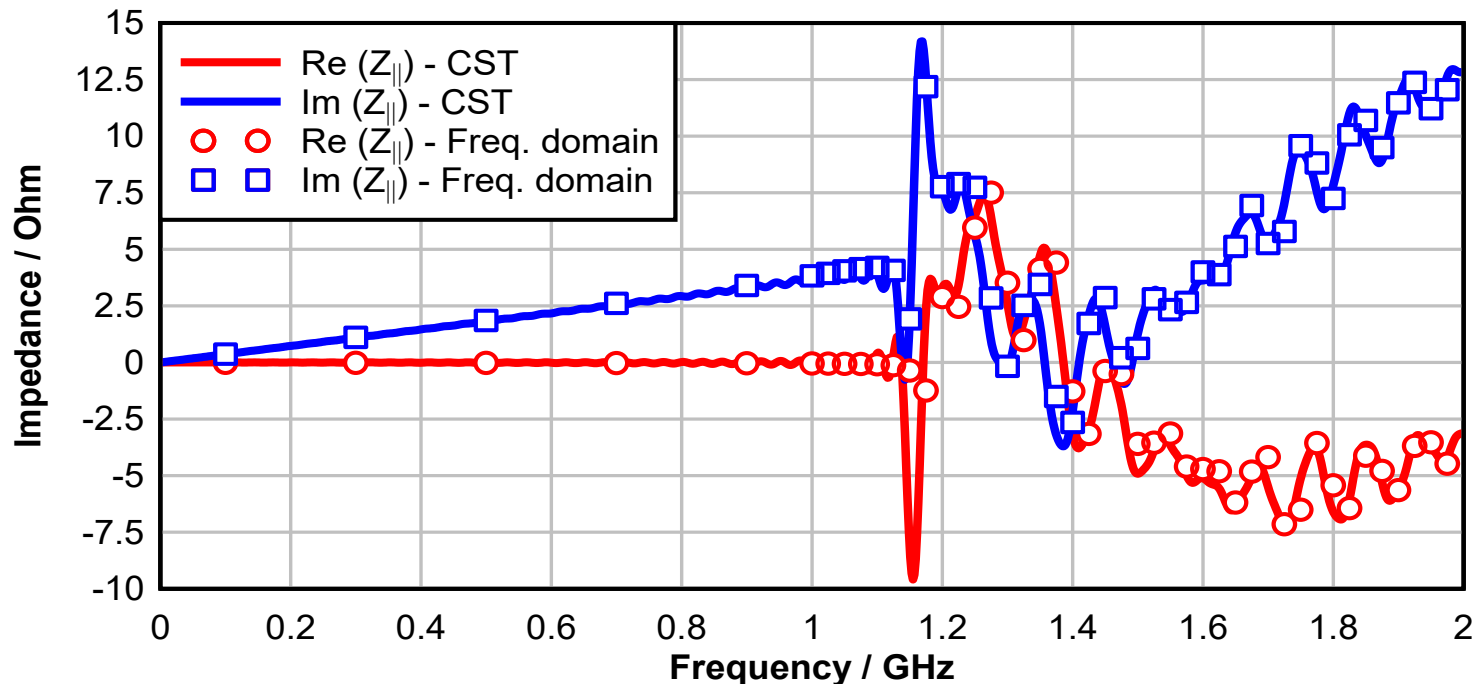
- Collimator example – impedance



$E_z - 1\text{GHz}$



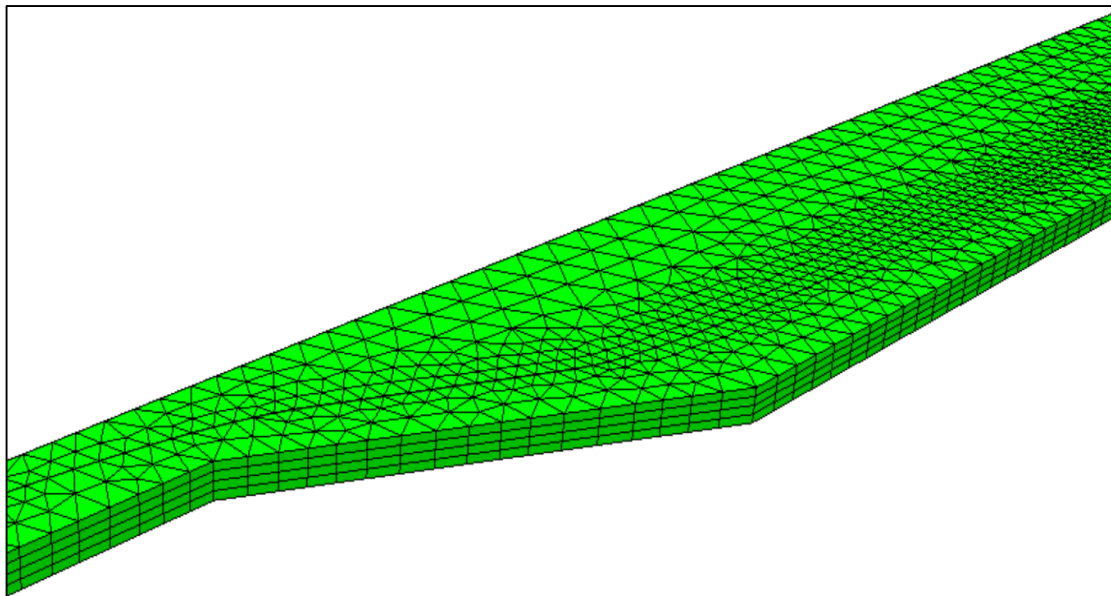
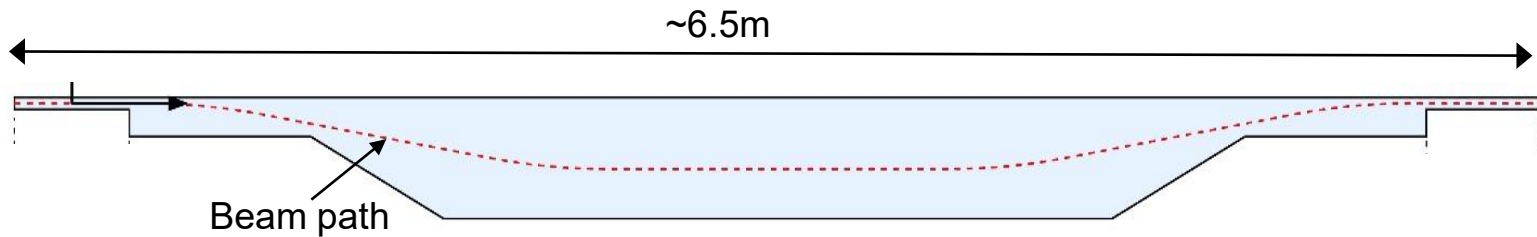
$E_z - 1.5\text{GHz}$



Comparison with
CST PS (time
domain)

Wakefields in the Frequency Domain

- Bunch compressor of XFEL, DESY



Prismatic element mesh:

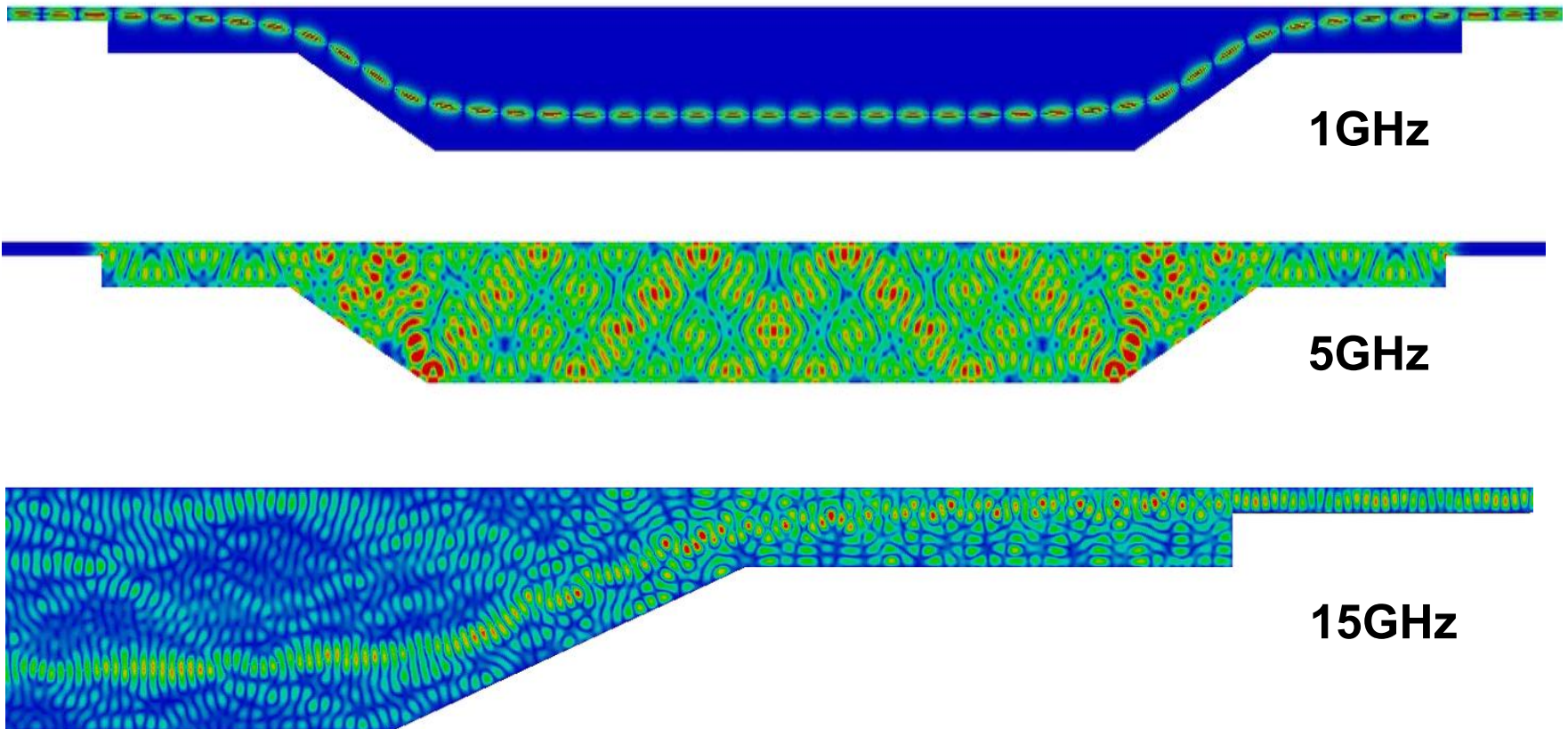
$\Delta \approx 5\text{mm}$

600k cells

4th order FEM

Wakefields in the Frequency Domain

- Bunch compressor of XFEL, DESY



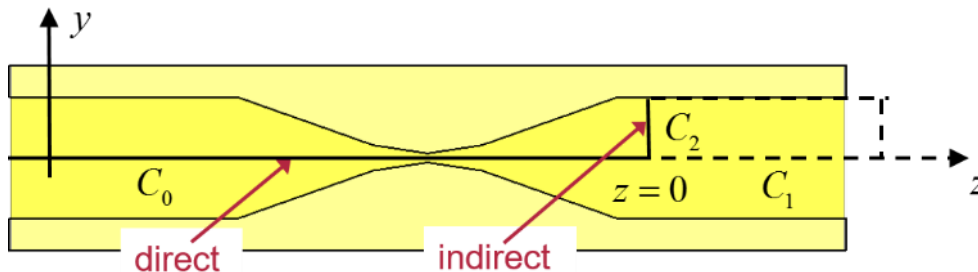
Summary & Conclusions

- **Wakefield codes in the time domain**
 - Moving window – not always applicable
 - Conformal boundary techniques – not always accurate
 - Resistive wall wakefields – ADE, expensive convolution
 - Electrically large structures – numerical efficiency
 - Long wake transients – numerical efficiency
- **CSR wakes**
 - CSR DG code using DG in the time domain
- **Frequency domain approach**
 - May be filling the gap for a number of applications
 - Difficult for very high frequency problems
 - Efficient solvers for large system of equations are needed

Thank You for your attention

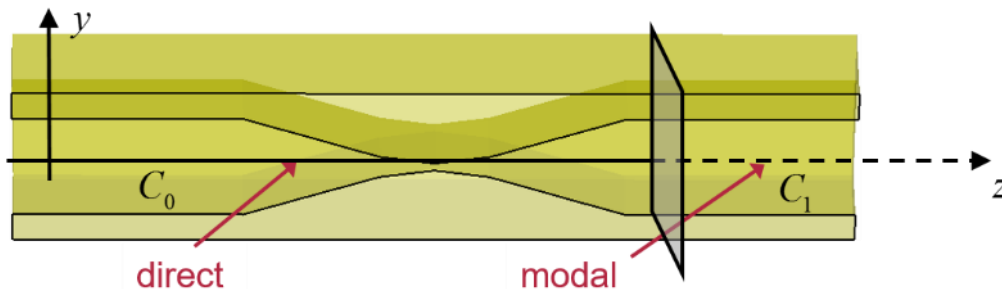
Indirect integration

- Indirect integration based on wake path deformation



- Weiland 1983, Napoly 1993
- A. Henke, W. Bruns, EPAC'06

- Indirect integration using modal decomposition

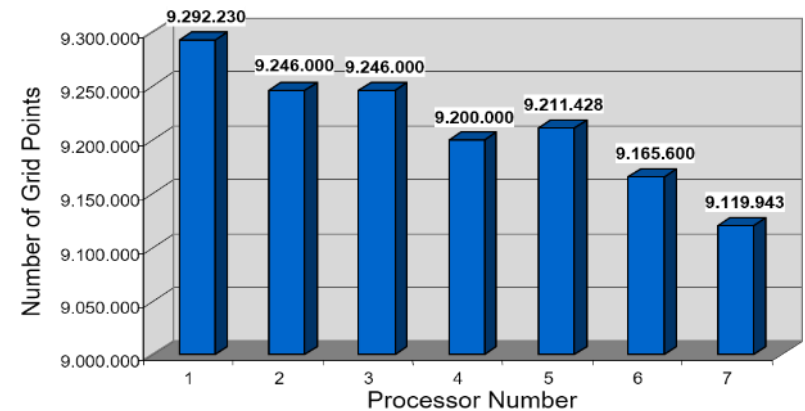
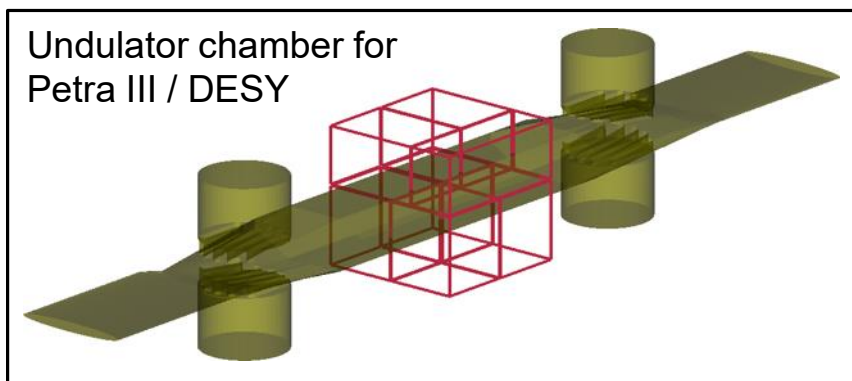
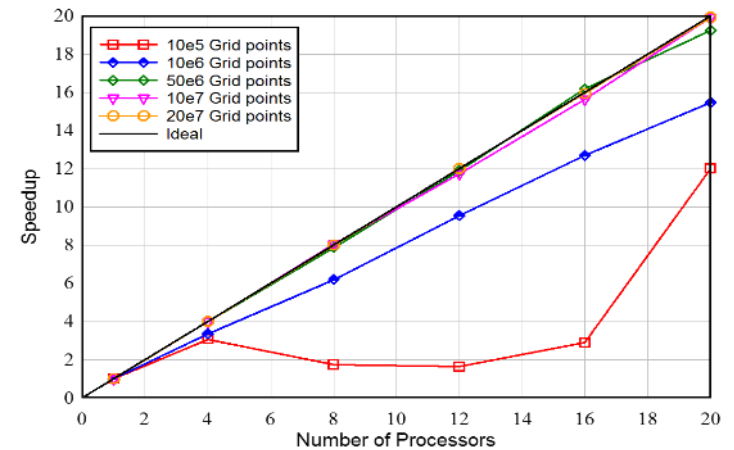
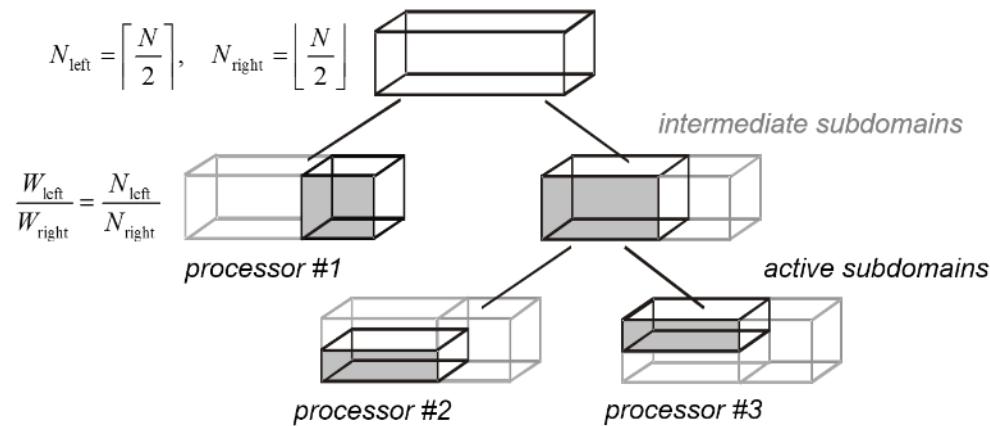


- I. Zagorodnov, PRSTAB 2006
- X. Dong, E. Gjonaj, ICAP'06

$$W_z(s) = -\frac{1}{Q} \int_{-\infty}^{\infty} dz E_z(z, t = \frac{z+s}{c}) = -\frac{1}{Q} \int_{C_0} dz E_z(z, t = \frac{z+s}{c}) - \frac{1}{Q} \sum_n e_z^n(x, y) W_n(s)$$

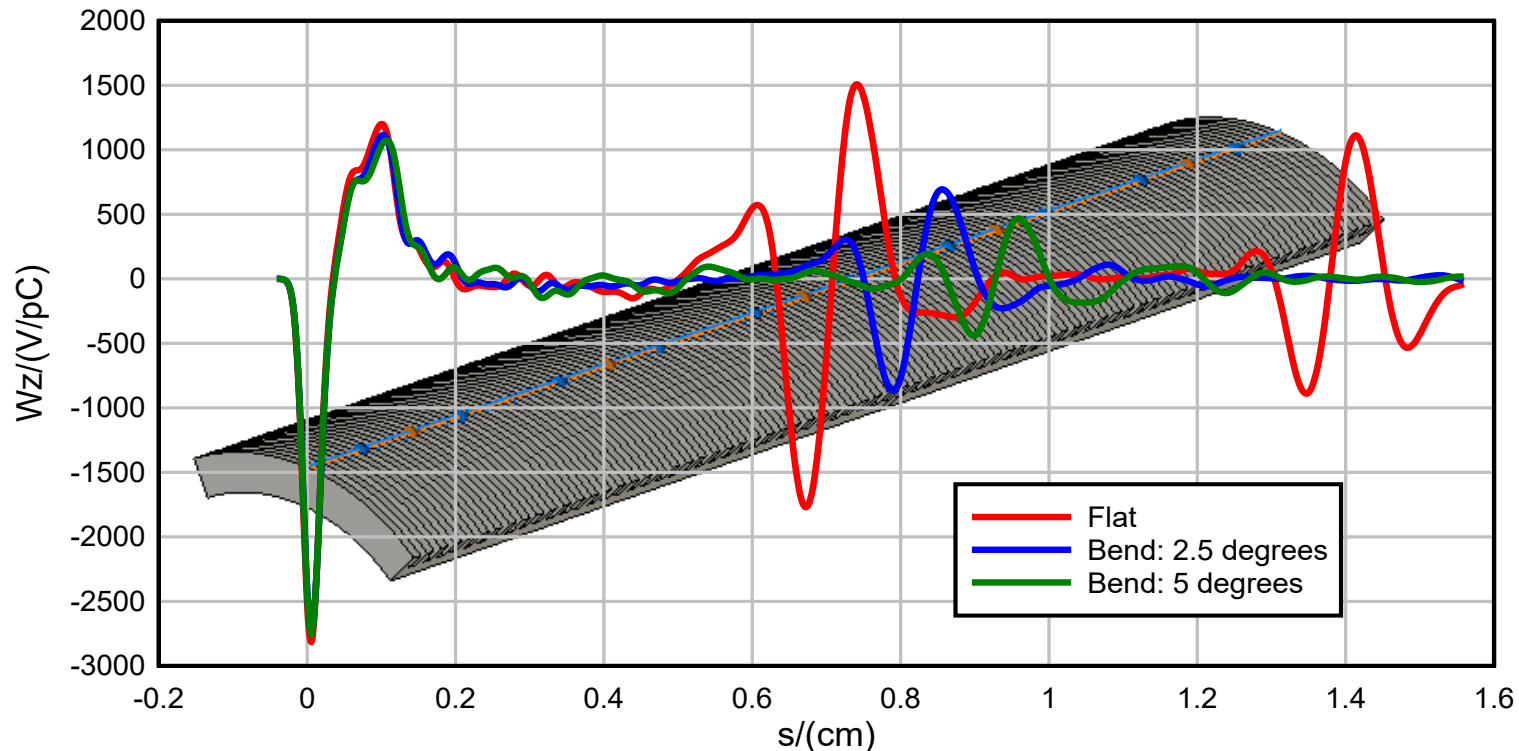
Parallelization

- Domain decomposition approach on large HPC-Clusters



- Single plate dechirper (with Bane, Stupakov)

Bend dechirper to minimize side end reflections



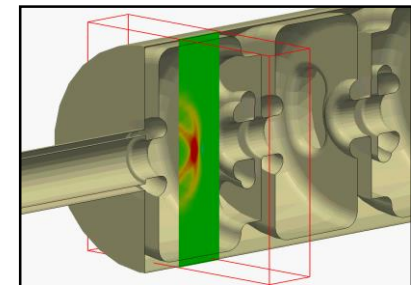
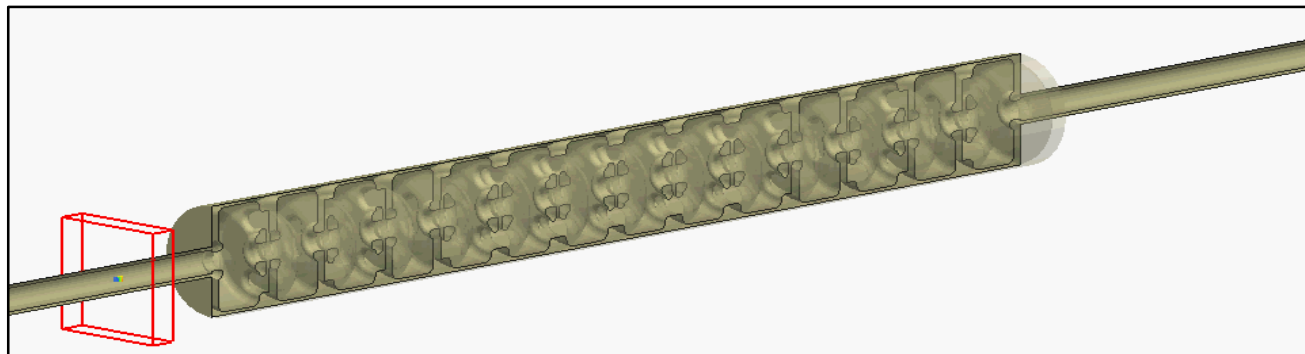
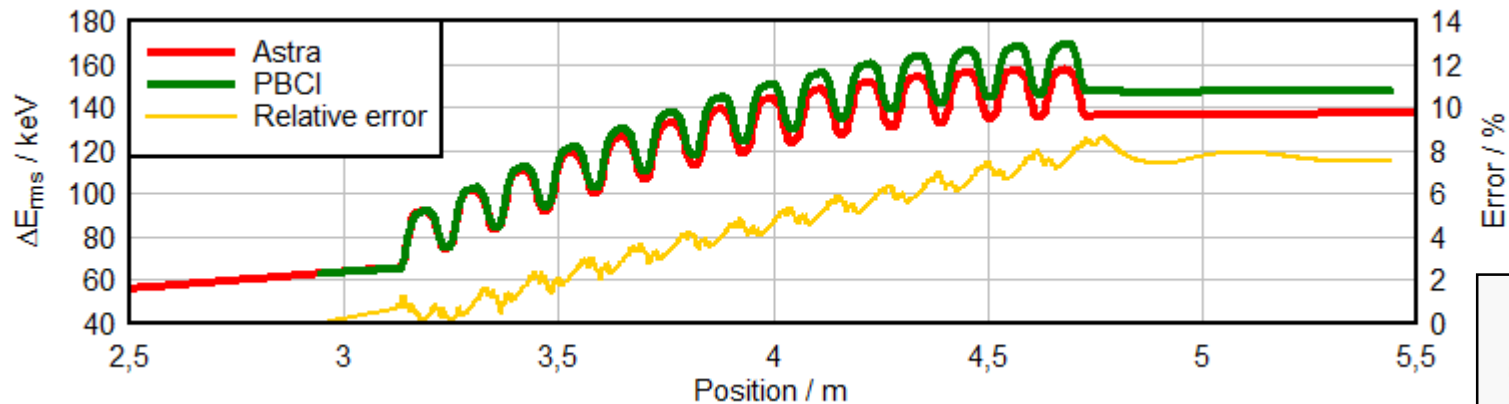
Wakefield simulation in the time domain

- Different approaches for the solution of Maxwell's equations
 - Fixed window codes
 - ABCI – FDTD
 - CST Particle Studio – FIT
 - Tech-X VSim – FDTD
 - T3P (SLAC) – high order FEM
 - GdfidL – FDTD/FIT
 - ...
 - Moving window and dispersion-free codes
 - GdfidL – FDTD/FIT
 - Echo2D, Echo3D (DESY) – FDTD/FIT
 - Parallel Beam Cavity Interaction PBCI (TEMF) – FDTD/FIT
 - ...

Particle tracking with wakefields

- Incorporate a particle tracker into the wakefield code

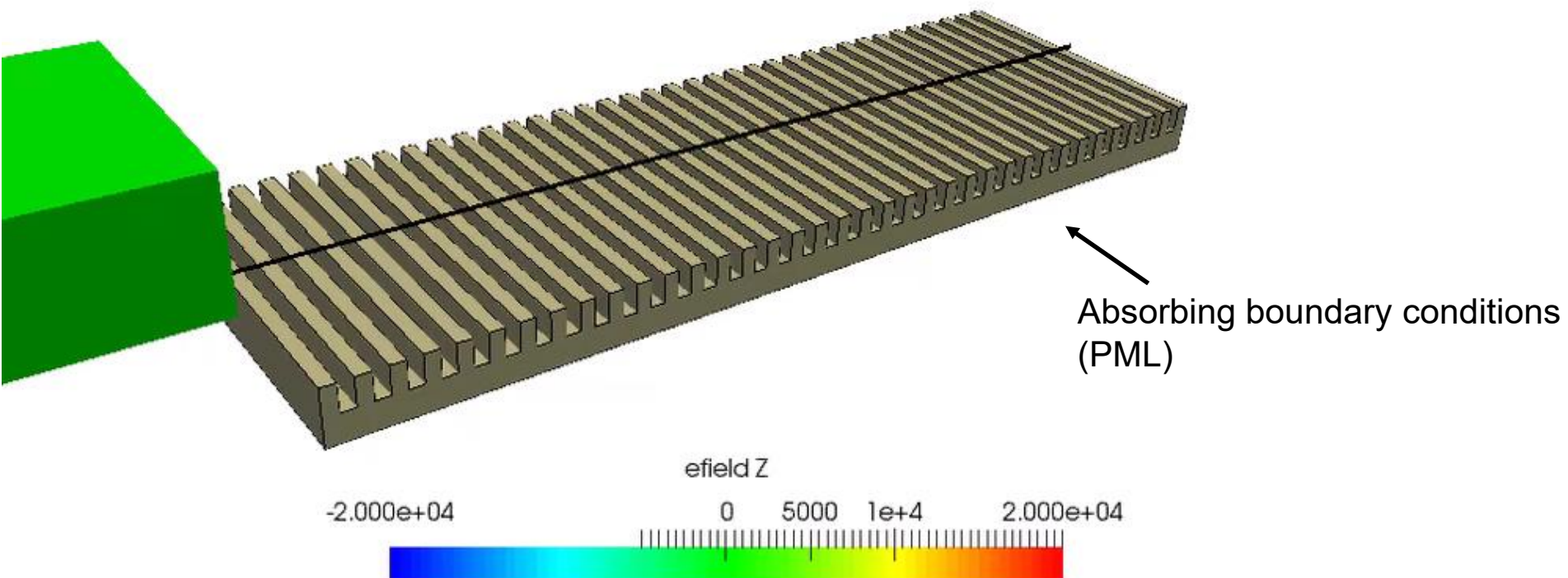
Influence of geometrical wakefields in the booster of the XFEL injector (PITZ, DESY)



Pure space-charge vs.
space-charge +
wakefield simulation

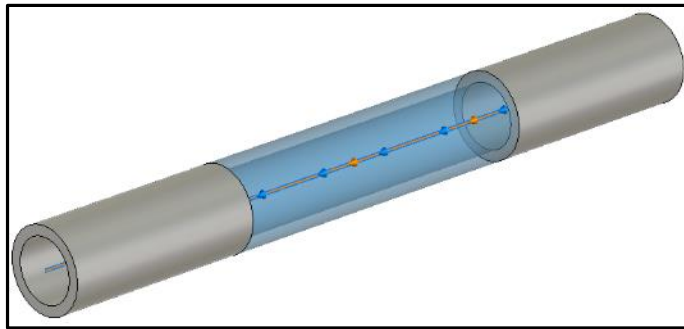
Open structures

- Single plate dechirper (with Bane, Stupakov)

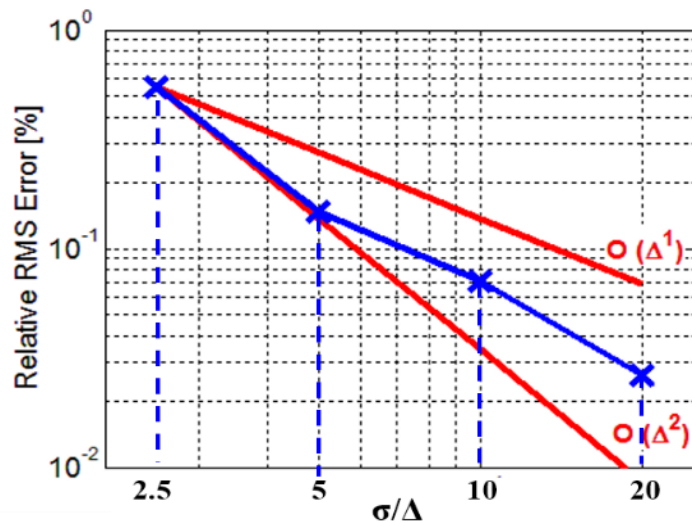


Resistive wall wakefields

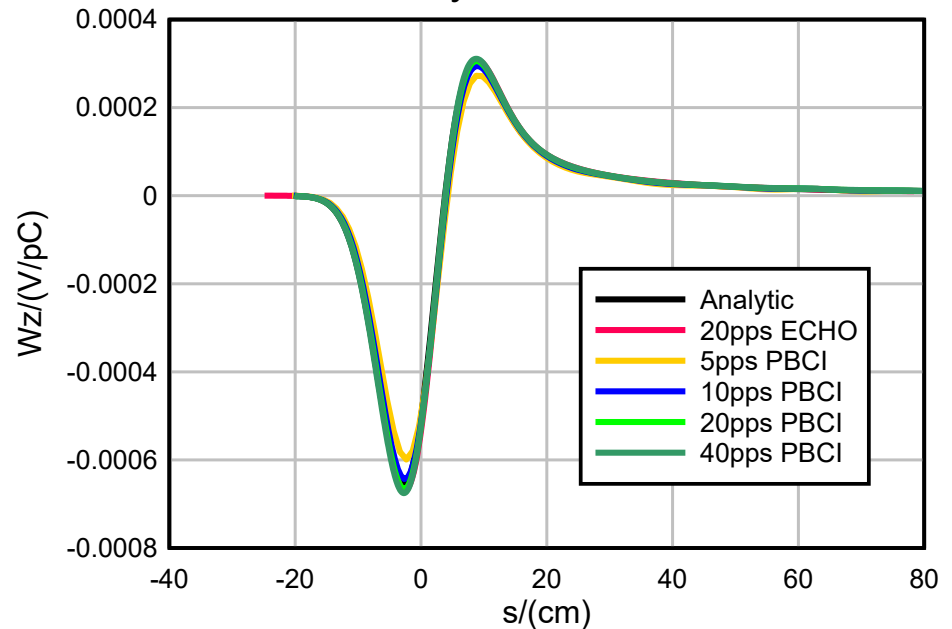
- Approximation of curved PEC boundaries



TiAl round pipe, $L=60\text{cm}$, $R=6\text{cm}$
Bunch: $\sigma=5\text{cm}$



Comparison with ECHO2D and analytical solution

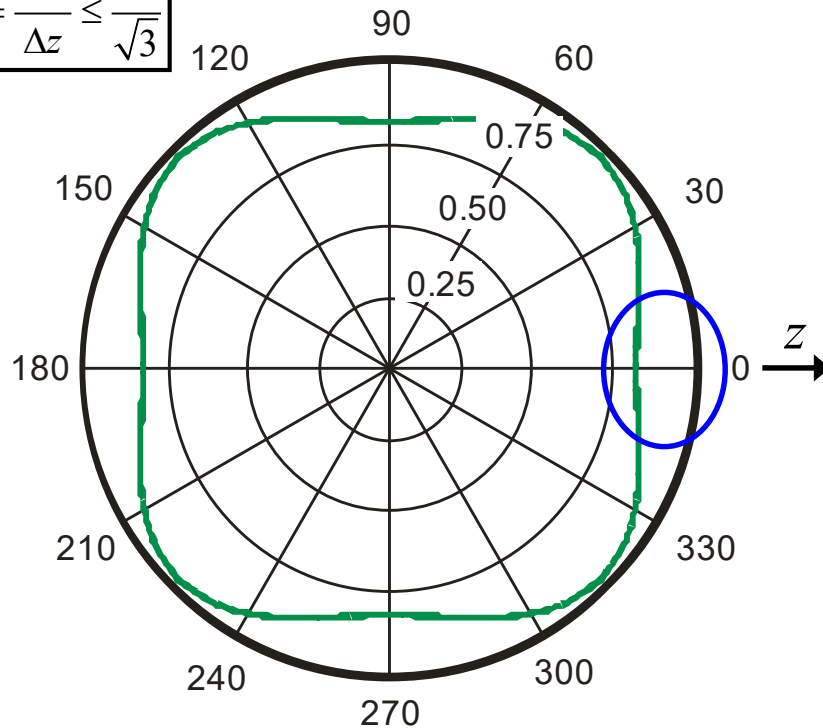


Dispersion-free methods

- Exact propagation in z-direction by splitting of the FDTD operators

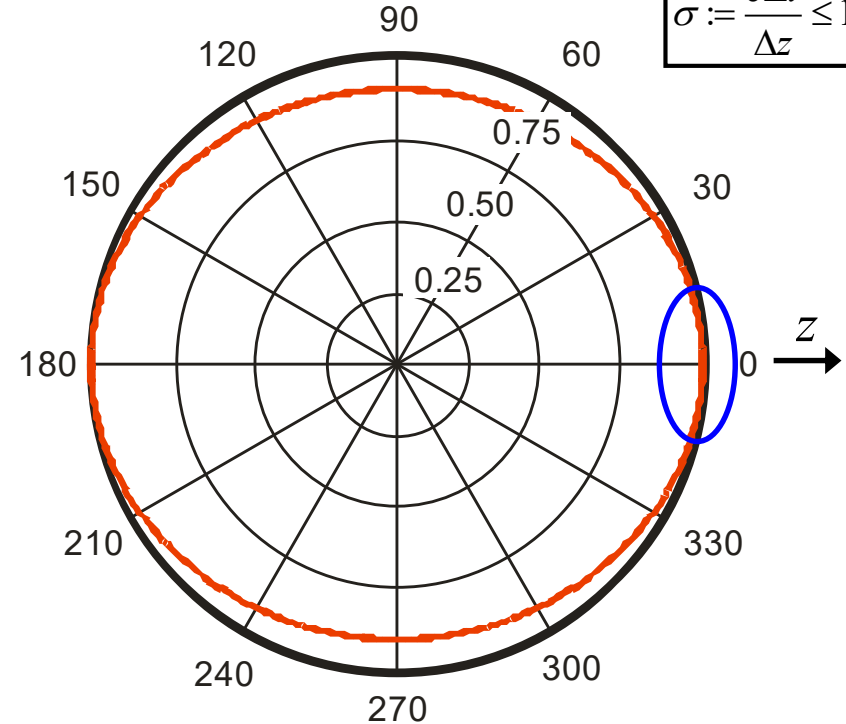
FDTD

$$\sigma := \frac{c\Delta t}{\Delta z} \leq \frac{1}{\sqrt{3}}$$



PBCI / LT

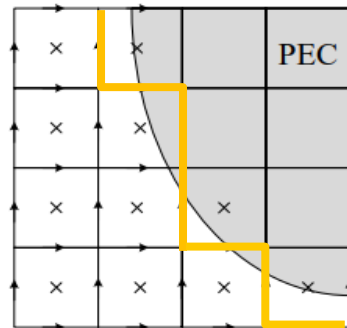
$$\sigma := \frac{c\Delta t}{\Delta z} \leq 1$$



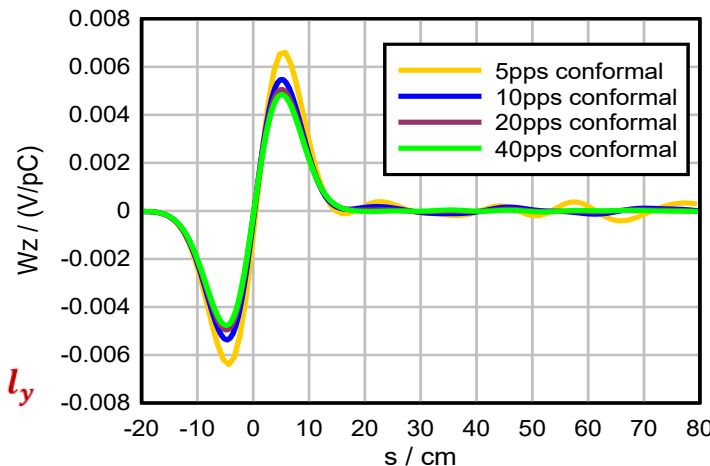
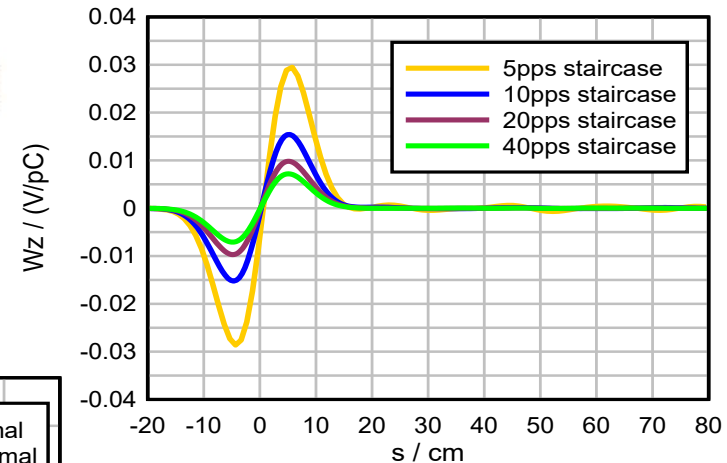
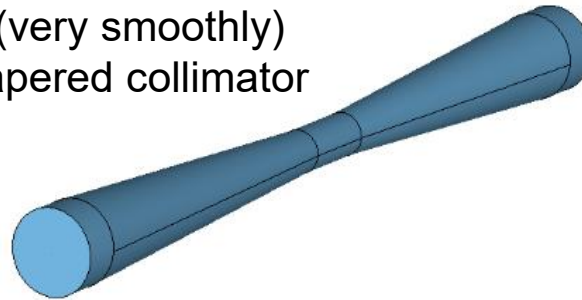
Geometry approximation

- Conformal approximation of curved boundaries
 - Local modification of discrete Faraday's law (Day&Mittra, Thoma)

staircase



(very smoothly)
tapered collimator



Up to 40 grid points per bunch length needed for good accuracy

